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# ORTHOGONAL TRANSFORMS FOR ELECTROCADIOGRAPHIC DATA COMPRESSION: A COMPARATIVE STUDY

التحويلات المتعامدة لاختزال بيانات رسم القلب الكهربي: دراسسة مقارنسسة

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الخلاصة . في هذا البحث تم مقارنة أداء ست خوارزمات للتحويلات المتعامدة في اختزال بيانات رسم القلب الكوبي وهي : تحويل فورير ، تحويل الجيب ، تحويل الجيب تمام، تحويل هارتلي، تحويل والش وتحويل هار الكوبى وهلى . تحويل توريز ، تحويل محبيب ، تحويل محبيب تسم. تحريل سارسي تحريل وقد أظهرت النتائج أن وقد استخدم لذلك عدد من التسجيلات الحية لرسم القلب لحالات عادية وأخرى غير عادية ، وقد أظهرت النتائج أن تحويل هارتلى بتمبر بعقدرته على اخترال البيانات مع تواجد أقل سبة مثوية للخطأ بين الاشارة الأصلية والاشارة التي تم استعادتها حيث تم مقارنة معامل الأدا، ( النسبة المثوية لجنر متوسط مجموع مربعات الغروق RMS ) الاكليبيكية للاشارة

ABSTRACT- Discrete forms of the Fourier, Sine, Cosine, Hartley, Walsh, and Haar transforms are examined for their performance to compress the electrocardiographic (ECG) signals. To compare their effectiveness in accomplishing this goal, the percent RMS difference resulting for each transformation method at various compression ratios is computed for records of normal and abnormal cases. Based on this comaprison, it is found that the Hartley transform is most advantageous for compresssing the ECG signal as it provides higher compression ratios and lower percent RMS differences compared to the other five transforms.

#### I. INTRODUCTION

Orthogonal transforms are one of the powerful techniques for data compression. The compression procedure involves preprocessing the input signal by means of a linear orthogononal transform such as Fourier, Karhunen-Loeve, Walsh, Haar, ..., etc., and properly encoding the transformed output (expansion coefficients) and reducing the amount of data needed to adequately represent the original signal. Upon signal reconstruction, an inverse transformation is performed and the original signal is recovered with a certain degree of error.

Many discrete orthogonal transforms have been employed to electrocardiographic (ECG) data compression. The discrete Cosine transform (CT), the Haar transform (HT), and the Karhunen-Loeve transform (KLT) have been applied, resulting in a compression ratio (CR) of 3:1 (1). Dual application of the KLT [2] to a vector lead ECG (X, Y, and 2 leads of the vectorcardiogram in the Frank coordinate system) partitioned into a P wave and a QRST segment have resulted in a CR of 12 : 1. Use of Fourier transform (FT) in two-lead ECG data compression has also been reported [3] giving a CR greater than 7,4. Discussion on the employment of the fast Walsh transform in ECG signals is given in (4) and further study is reported in (5) resulting a CR of 8. Application of the discrete Hartley transform (DHT) has been reported in [6] and resulted in a CR of 12 :1. However, due to the diverse procedures that have been employed in each application, it would be improper

to compare these methods in absolute terms of values given for CR and percentage error as they have been obtained under different conditions of sampling rate, presicion, and noise level, and their performance have been evaluated using different indexes of error. To arrive at an exact comparison, a large set of ECG's from a common database needs to be processed by all ECG compression techniques and their performance needs to be evaluated with a common measure of "goodness" [7].

In the present paper, a comparative study is presented to evaluate the performance of six orthogonal transforms. These are: Fourier, Cosine, Sine, Hartley, Walsh and Haar transforms. The study was conducted using records of human ECG's for normal and abnormal cases. In the following section of this paper, we present the theoretical basis behind data compression via orthogonal transforms and a general introduction to each transform is given. The third section includes a description to the database used for evaluation. The fourth section is devoted to the description of the compression scheme using orthogonal transforms and the implementation of each transform. Evaluation of the performance of the transforms techniques and some representative results are presented in the fifth section for a number of normal and abnormal cases by reconstructing real ECG signals.

#### II. ORTHOGONAL TRANSFORMS FOR DATA COMPRESSION

Consider a discrete data vector  $X^T = \{x_0, x_1, \dots, x_{N-1}\}$  and an orthogonal transform  $P = \{\psi_0, \psi_1, \dots, \psi_{N-1}\}^T$ , where  $\psi_0, \psi_1, \dots, \psi_{N-1}\}$  are orthogonal real-valued vectors, i.e. vectors having the property  $\psi_j^T \psi_j = \delta_{i,j}$  (Kronecker delta) or  $P^T P = I$  (unit  $\aleph \times \aleph$  matrix). Applying P upon X gives transformed vector components  $y_i = \psi_i^T X$  or in matrix form

$$Y = PX, Y^{T} = \{y_{0}, y_{1}, \dots, y_{N-1}\}$$
 (1)

Premultiplying (1) by  $P^T$  and using the orthogonality of P gives the inverse transform

$$X = P^{T} Y = \sum_{i=0}^{N-1} y_{i} \psi_{i} \qquad (P^{T} = P^{-1})$$
 (2)

Clearly, P and P<sup>-1</sup> are linear inverse transforms. The basic problem is to select a family of vector functions  $\psi_{\circ}$ ,  $\psi_{i}$ , ...,  $\psi_{N-1}$  such that the signal X can be reconstructed or estimated using a smaller number of terms  $y_{i}\psi_{i}$ ,  $i=0,1,\ldots,M-1$ . This implies that only M < N data values  $y_{i}$  have to be used or transmitted for reconstructing X as

$$X = \sum_{i=0}^{M-1} y_i \psi_i + \sum_{i=M}^{N-1} C_i \psi_i$$
 (3)

where  $c_i$  are preselected constants. Using fewer data points or transform coefficients  $y_i$ , we have what is known as "data compression" which allows to transmit the information signal X approximately with reduced information rate, a result which is mostly important in practice.

#### II.1 The Optimal Transform

Since we are interested in dropping some of the coefficients, we may also want the error (mean-square) in reconstruction of X, due to dropping of some coefficients be the least. Also we would like signal energy to be compacted into as few coefficients as possible. Fortunately, there is one transform that satisfies this criterion, and it is known as the Hotelling transform or Karhunen-Loeve transform (KLT) [8]. The optimum transform can be computed from the covariance matrix C

$$C_{i} = E \{(X - E(X)), (X - E(X))^{T}\}\$$
 (4)

where E is the statistical expectation and superscript  $\tau$  denotes transpose. Indeed, the mean square error V is given by

$$V = E \{ (X \sim E(X))^{T} (X \sim E(X)) \}$$

$$= \sum_{i=M}^{N-1} E \{ (Y_{i} - C_{i})^{2} \}$$
(5)

and the optimality conditions  $\partial V/\partial c_i = 0$ ,  $\partial V/\partial \psi_i = 0$  subject to  $\psi_i \psi_i^T = 1$  give, respectively:

$$c_{i} = E [ y_{i} ] = \psi_{i}^{T} E(X),$$

$$\sum_{x} \psi_{i} = \lambda_{i} \psi_{i}$$
(6)

where  $\lambda_i$  are the Lagrange coefficients for the constraints  $\psi_i^T\psi_i=1$ . This shows that  $\psi_i$  is an eigenvector of  $\Sigma_{\mathbf{x}}$  and  $\lambda_i$  is the corresponding eigenvalue. Now from Y = PX we obtain

$$\Sigma_{y} = P \Sigma_{x} P^{-1} = P \Sigma_{x} P^{T} = \text{diag} \{\lambda_{0}, \lambda_{1}, \ldots, \lambda_{N-1}\}$$
 (7)

i.e.  $\Sigma_{\mathbf{y}}$  is diagonal which implies that the components of  $\mathbf{y}_{i}$  are uncorrelated.

The minimum value  $V_{o}$  is obtained by substituting (6) into (5), i.e.

$$V_{o} = \sum_{i=M}^{N-1} \psi_{i}^{T} \Sigma_{N} \psi_{i} = \sum_{i=M}^{N-1} \psi_{i}^{T} \lambda_{i} \psi_{i} = \sum_{i=M}^{N-1} \lambda_{i}$$
(8)

This shows that V is equal to the sum of the eigenvalues corresponding to the data components  $y_i$  which are discarded. Thus in practice, one must keep the M data components having the largest eigenvalues. The discarded data components are replaced by the constants  $c_i = \psi^T E(x)$  which can be reduced to zero by preprocessing the given data such that E(x) = 0. This data compression criterion for selecting the set of  $y_i$  components with the largest variances is known as "variance data compression criterion". Note that for all other orthogonal transforms the covariance matrix  $\Sigma_y$  is not diagonal (9).

An alternative way of expressing the above is to use a set of basis vectors  $\phi_1, \ldots, \phi_N$  which are orthonormal and then represent X as a linear sum of these basis vectors, i.e.,

$$X = \sum_{i=1}^{N} a_i \phi_i \tag{9}$$

It is well known [9] that the optimum  $a_{iN}$  which minimize the error between X and its representation (i.e.,  $\sum_{i} a_i \phi_i$ ) are given by

$$\mathbf{a}_{i} = \mathbf{X}^{\mathsf{T}} \boldsymbol{\phi}_{i} \tag{10}$$

Now instead of storing signal samples, we store  $\{a_i\}$  only. In order to achieve the maximum degree of compression, we would like to store as few  $\{a_i\}$  as possible, without a significant error in the reconstruction of X. The basis vectors which compact maximum energy in the fewest  $\{a_i\}$  turn out to be the eigenvectors of the covariance matrix  $C_x$ . Thus the KLT provides us with the optimum basis functions for representing a signal.

Although the optimum transformation is explicitly known, its use in practice presents many problems [7]. First of all, the computational time needed to calculate the KLT basis vectors are based on determining the eigenvalues and corresponding eigenvectors of the covariance matrix of the original data, which can be a large symmetric matrix. Its implementation requires  $N^2$  multiplications by constants which may be complicated (i.e., not simple powers of 2 for easy binary arithmetic). The second problem is in the computation of the eigenvectors of  $C_{\rm x}$ . In many cases this matrix  $C_{\rm x}$  turns out to be singular, and then some eigenvectors cannot be uniquely defined. The lengthy processing requirement of the KLT has led to the use of suboptimum transforms with fast algorithms [7].

#### II.2 Suboptimum Transforms

Many other transforms have been invented which produce less correlated coefficients than the signal itself and which are easier to implement. Some of the popular transforms are:

# 1) Discrete Fourier Transform (DFT)

where

$$a_{ij} = \frac{1}{\sqrt{N}} \exp \left[-2\pi \sqrt{-1} - (ij)\right]$$
 (11)

# 2) Discrete Cosine Transform (DCT)

$$a_{ij} = \frac{2K(i)}{\sqrt{N}} \cos \left[ (2j + 1)i\pi/2N \right]$$
 (12)

$$K(i) = \begin{cases} 1/\sqrt{2}, & \text{for } i = 1\\ 1, & \text{for } i = 2, ..., N\\ 0. & \text{otherwise} \end{cases}$$

# 3) Discrete Sine Transform (DST)

$$a_{ij} = \sqrt{\frac{2}{N+1}}$$
  $\sin \left[ \pi(i+1)(j+1)/(N+1) \right]$  (13)

# 4) Discrete Hartley Transform (DHT)

$$a_{ij} = \cos(2\pi ij/N) + \sin(2\pi ij/N)$$
 (14)

# 5) Discrete Walsh Transform (DWT)

$$a_{ij} = \frac{1}{\sqrt{N}} (-1)^{b(i,j)}$$
 (15)

where 
$$b(i,j) = \sum_{l=0}^{k-1} i_l J_l$$

The terms  $\boldsymbol{\varepsilon}_{l}$  and  $\boldsymbol{j}_{l}$  are the bit states of the binary representations of i and j, respectively.

## 6) Discrete Haar Transform (HT)

$$a_{ij} = 1$$

$$= 2$$

$$for 0 < n < 1$$

$$for  $\frac{j-1}{2^{i}} \le n \le \frac{j-1/2}{2^{i}}$ 

$$= -2$$

$$for \frac{j-1/2}{2^{i}} \le n \le \frac{j-1/2}{2^{i}}$$

$$= 0$$

$$for \frac{j-1/2}{2^{i}} \le n \le \frac{j-1/2}{2^{i}}$$

$$= 0$$

$$for \frac{j-1/2}{2^{i}} \le n \le \frac{j-1/2}{2^{i}}$$

$$= 0$$

$$for \frac{j-1/2}{2^{i}} \le n \le \frac{j-1/2}{2^{i}}$$$$

The set of orthogonal functions may be any one of many different types. In the Fourier, Cosine, Sine and Hartley transforms, the orthogonal functions are discrete sampled sines and/or cosines as illustrated in Fig.1 for N = 16. In the Walsh and Haar transformations, the orthogonal functions are discrete sampled walsh functions (10) (Fig.1).

#### III. DATA ACQUISITION

35-sec Lead II ECG signals were recorded using a Philips ECG recorder type XV1503. The signals were separated into two classes, namely normal and abnormal. The abnormals showed old inferior infarction. The bandwidth of the signal was chosen to be 100 Hz. The output of the ECG recorder was then connected to a 12-bit analog-to-digital converter and sampled at a rate of 250 samples/sec. The digital data were transferred to an IBM PS2/80 computer by a program written in Basic.

#### IV. THE COMPRESSION SCHEME

IV.1 Preprocessing

The method of orthogonal transforms requires delineation of the QRS complexes. This is usually accomplished using a search procedure based on the first derivative of the signal. Once the QRS is detected, backward and forward searches are performed with a lower threshold to locate its beginning and end. Each ECG beat was represented by 256 samples. These samples were chosen such that the QRS complex and the T wave would always appear.

IV.2 Implementation

Fig.2 shows a block diagram of the compression scheme. The input to the scheme is the ECG signal X of length N samples. The forward transform yields the coefficients  $a_{ij}$ . The coefficients to be retained are lower order harmonics. The number of coefficients L to be retained to achieve a given compression ratio CR is given

$$L = N/CR \tag{17}$$

The higher order harmonics for each transform are discarded. For reconstruction, N-L zeros are added to the L coefficients and the resulting sequence is inverse transformed. The reconstructed waveforms are compared to the original signals by visual examination of plotted waveforms and also by a measure of goodness as described in the next subsection.

The obvious difference between these transforms and the KLT is the ease of implementation. All the above transforms are unitatry, i.e.,  $A^{T*} = A^{-1}$  (where  $A^{T*}$  is the complex conjugate of  $A^{T}$ ), and therefore inverse transformation which is done for reconstruction is as the transformation itself.

The discrete Fourier transform was implemented by using the fast Fourier transform technique [11]. Instead of the N<sup>2</sup> multiplications required for the computation of the DFT,  $2N \log_2 N$  multiplications and additions are required. The discrete cosine transform was implemented by a technique developed by Makhoul [12] using N  $\log_2 N$  real multiplications. The Sine transform was computed using a fast algorithm developed by Yip and Rao (13). The discrete Hartley transform was

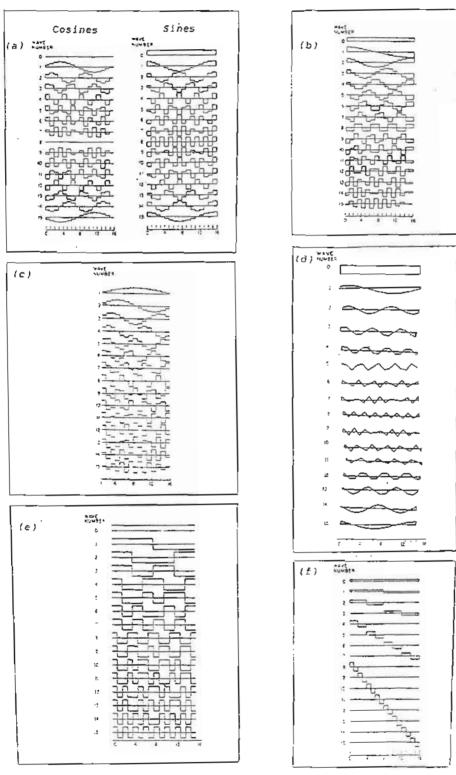


Fig.1 The basis orthogonal functions of the transforms used (N =16) (a) Fourier, (b) Cosine, (c) Sine, (d) Hartley, (e) Walsn, and (f) Haar.

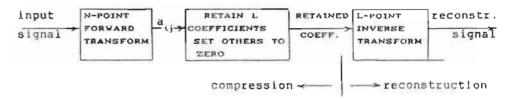


Fig. 2 Block Diagram of data compression scheme using orthogonal transforms

implemented using a radix-2 decimation-in-time fast Hartley algorithm [14]. The walsh transform is the simplest to implement, since it consists of  $\pm$  1's and, therefore, only additions and no multiplications are required. However, fast algorithms have been developed that can compute an N-point Walsh transform in N  $\log_2$  N rather than N² operations [15]. The Haar transform was computed using a fast Haar transform algorithm [16]

#### IV.3 Performance Index

In order to assess the performance of the compression scheme, in addition to visual comparison, an index of performance was employed. It represents a measure of "goodness" of the reconstructed waveforms. This index is the percent RMS difference (PRD), which is computed as

$$PRD = \left[ \begin{array}{c} \sum_{i=1}^{N} \left( X(i) - \hat{X}(i) \right)^{2} \\ \sum_{i=1}^{N} \left( X(i) \right)^{2} \end{array} \right]^{1/2}$$
 (18)

where X and  $\bar{X}$  are samples of the original and reconstructed data sequence.

Using this method, we can either compute the error between the original and reconstructed waveforms for a specified compression ratio or obtain the permissible ratio for the specified error limit.

# V. RESULTS

Fig.3 shows a typical normal ECG record. Fig.4 depicts its six spectra corresponding to the six orthogonal transforms.

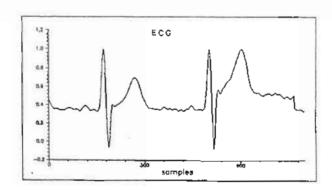


Fig.3 A typical ECG signal (normal record)

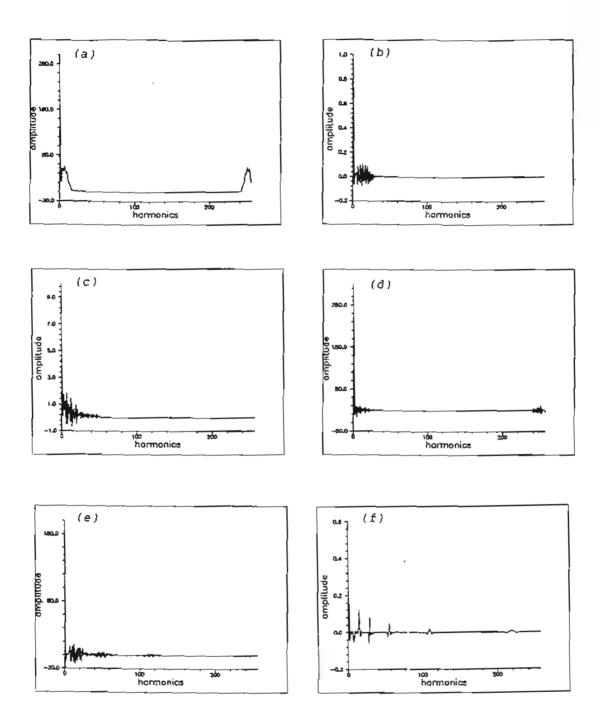


Fig.4 Spectra of the six orthogonal transforms of the signal of Fig.3. (a) Fourier, (b) Cosine, (c) Sine, (d) Hartley, (e) Walsh, and (f) Haar.

TABLE I

Percent RMS Differences for various orthogonal transforms (normal cases)

Compression ratio CR	Coeff. retain. L	APRD						
		DFT	DCT	DST	DHT	DWT	HT	
5.1	20	36.60	26.16	6.37	1.97	9.07	9.49	
6.4	30	36.92	36.07	8.67	2.52	11.27	11.43	
8.6	40	37.97	36.15	11.19	3.65	13.38	15.06	
12.8	50	46.70	49.36	20.30	13.97	23,30	28.24	

Figs.(5,6) show reconstructed waveforms (dotted line) superimposed over the original signal (continuous line) for each transform using compression ratios CR = 6.4, 12.8 respectively. Fig. 7 shows the results of compressing and reconstructing an abnormal record using CR = 8.6. As seen from the figures, as the CR value increases, the more distortion in the shape of the beat occurs. The figures also show that the two transforms resulting in the lowest distortion are the Sine and Hartley transforms. The other four transforms produced major variations even in the level of the isoelectric line of the ECG signal and, therefore, resulting changes in the magnitude of the R-wave with respect to the isoelectric line. Inspection of the figures of the Hartley transform shows that the information lost as a consequence of the data compression is not diagnostically significant: the Sine transform yielded more oscillatory variations.

The performance indexes described in Section IV.3 were computed for two hundred ECG beats of which half were normal and the other half were abnormal. Table I summarizes the average PRD's (APRD) for the different transforms using CR = 5.1, 6.4, 8.6, and 12.8 in case of normal records. The APRD values obtained for the abnormal signals are summarized in Table II. The results tabulated confirm the visual comparison. The Hartley transform possesses higher capability to restore the original signals with the lowest APRD's value.

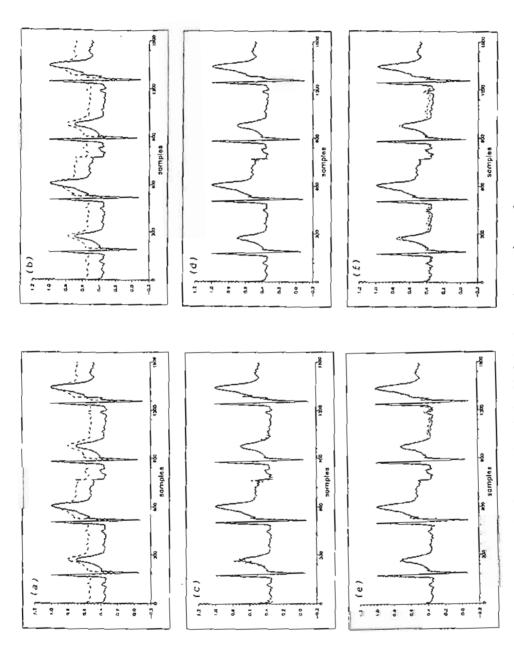
TABLE II

Percent RMS Differences for various orthogonal transforms
(abnormal cases)

Compression ratio CR	Coeff. retain. L	APRD						
		DFT	DCT	DST	DHT	DWT	HT	
5.1	20	34.65	29.96	6.45	2.03	9.87	8.47	
6.4	30	36.97	38.84	8.35	3.44	13.56	12.32	
8.6	40	37.24	45.84	12.58	5.89	23.39	28.82	
12.8	50	45.75	50.12	22.47	14.95	25.34	29.44	

#### VI. CONCLUSION

In this study, six orthogonal transforms were implemented and evaluated for ECG data compression. The six transforms are: Fourier, Cosine, Sine, Hartley, Walsh, and Haar. Implementations of these transforms were carried out using fast techniques derived from their discrete versions.



Flq.5 Reconstructed ECG data (- - -) superimposed over the original signal (---) for a normal subject (CR = 6.4), (a) Fourier, (b) Cosine, (c) Sine, (d) Hartley, (e) Walsh, and (f) Haar.

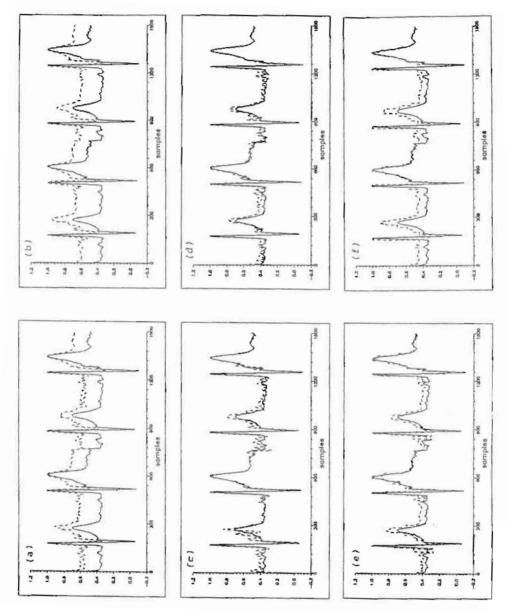


Fig.6 Reconstructed ECG data (- --) superimposed over the original signal (--) for a normal subject (CR = 12.8). (a) Fourler, (b) Cosine, (c) Sine, (d) Hartley, (e) Walsh, and (f) Haar.

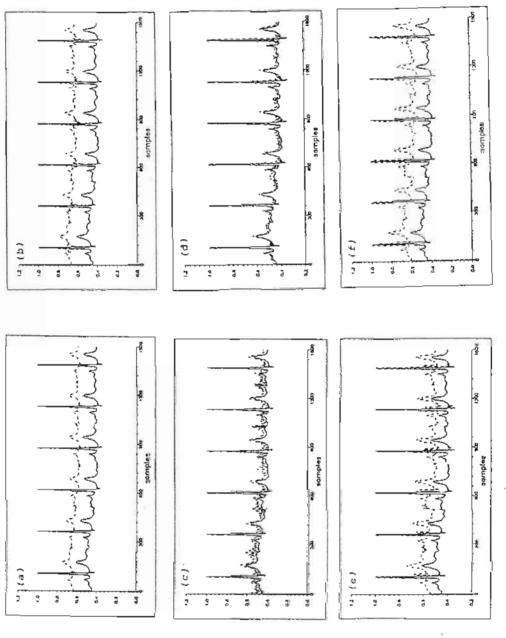


Fig.7 Recommittuected ECG data (- - -) superimposed over the original signal (---) for a subject having old inferior infarction (CR = 8.6).(a) Fourier, (b) Cosine, (c) Sine, (d) Haxtley, (e) Walsh, and (f) Haar.

The performance of the six orthogonal transformations was investigated in terms of the percent RMS difference PRD between the reconstructed and the original waveforms of records of normal and abnormal cases. This has demonstrated that the Hartley transform is the most efficient and accurate transformation that can be used safely for ECG data compression. It allows the restoration of the compressed signal with the lowest PRD value for the same compression ratio.

# References

Ahmed, P. J. Milne, and S. G. Harris, ardiographic data compression via orthogonal N. "Electrocardiographic transforms," IEEE trans. Biomed. Eng., vol. BME-22, no. 6, pp. 484-487, Nov. 1975.

[2] M. E. Womble, J. S. Halliday, S. K. Mitter, M. C. Lancaster and J. H. Triebwasser, "Data compression for storage and transmitting ECG's/ VCG's," Proc. IEEE, vol. 65, no. 5, pp.

702-706, May 1977. [3] B. R. S. Reddy and I. S. N. Murthy, "ECG data compression using Fourier descriptors," IEEE trans. Biomed. Eng., vol. BME-33, no. 4, pp. 428-438, Apr. 1986.

[4] W. S. Kuklinski, "Fast walsh transform data-compression algorithm: e.c.g. applications, " Med. & Biol. Eng. & Comput.,

vol. 21, pp. 465-472, July 1983. [5] G. P. Frangakis, G, Papakonstantinou and S. G. Tzafestas, "A fast walsh transform-based data compression multiprocessor system: applications to ECG signals," Math. Comput. simulation,

vol. 27, pp. 491-502, 1985. [6] F. E. Z. Abou-Chadi, "Fast Hartley transform data-compression algorithm: ECG applications," Submitted for publication in the

Mansoura Eng. Journal (MEJ).

[7] S. M. S. Jalaleddine, C. G. Hutchens, R. D. Strattan, and W. A. Coberly, "ECG data compression techniques- A unified approach," IEEE trans. Biomed. Eng., vol. BME-37, no. 4, pp. unified 329-343, Apr. 1990.

[8] A. N. Netravali and J. O. Limb, "Picture coding: a review,"

Proc. IEEE, vol. 68, no. 3, pp. 366-406, March 1980.
[9] D. F. Elliott and K. R. Rao. Fast transforms: algorithms, analyses, applications. Academic Press, New York, 1982.

(10) K. B. Beauchamp. Walsh functions and their applications. Academic Press, New York, 1975.

[11] D. M. Monro. Fortran 77. Butler and Tanner Ltd, 1982.

[12] J. Makhoul, "A fast cosine transform in one and two dimensions," IEEE trans. Acoust, Speech, and Signal Proces., vol. ASSP-28, no. 1, pp. 27-34, Feb. 1980.

[13] P. Yip and K. R. Rao, "A fast computational algorithm for the discrete sine transform," IEEE trans. Commun., vol. COM-28,

no. 2, pp. 304-307, Feb. 1980. [14] H. V. Sorensen, D. L. Jones, C. S. Burru, and Heideman, "On computing the the discrete Hartley transform," IEEE trans. Acoust., Speech, and Signal Proces., vol. ASSP-33, no. 4, pp. 1231-1238, Oct. 1985.

[15] R. D. Brown, "A recursive algorithm for sequency-ordered fast Walsh transforms'" IEEE trans. Comp., vol.C-26, pp. 819-823,

[16] N. Ahmed, T. Nataragan, and K. R. Rao, "Cooley-Tukey type algorithm for the Haar transform," Electron. Lett., no.9, pp. 276-278, 1976.